

Adaptive Signal Processing and System Theory

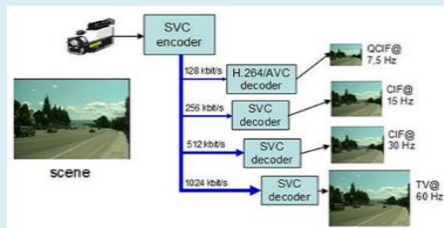
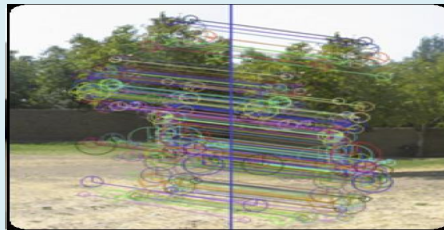
(#2: The z-transform)



2023 Spring

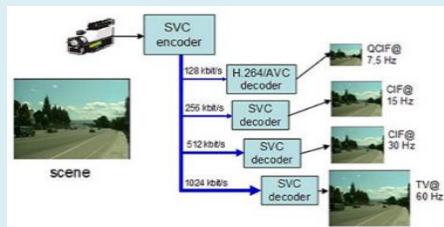
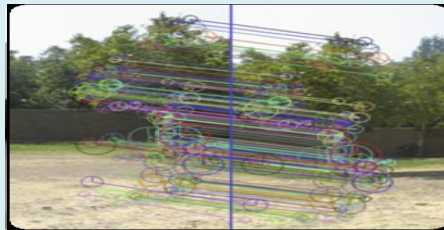
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Contents

- Introduction of the z-Transform
- Relationship of Properties of Seq. to Its z-Transform
- Z-Transform for LTI system



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- ❖ z-Transform representation of a sequence
 - Fourier transform does not converge for all sequences.
 - ➡ Generalization of Fourier Transform (FT)
 - In analytical problems, the z-transform notation is often more convenient than FT notation.

z-Transform (1)

❖ Fourier transform (FT)

$$X(e^{j\omega}) = \sum x[n]e^{-j\omega n}$$

❖ z-Transform

$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n} \quad : \text{bilateral z-transform}$$

where z = a constant complex variable.

$$x[n] \xleftrightarrow{Z} X(Z)$$

▪ Unilateral z-transform

$$\mathcal{X}(z) = \sum_{n=0}^{\infty} x[n]z^{-n}$$

z-Transform (2)

- In general, $z = re^{j\omega}$ (since z is complex value,)

$$\begin{aligned} X(re^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n](re^{j\omega})^{-n} \\ &= \sum_{n=-\infty}^{\infty} (x[n]r^{-n})e^{-j\omega n} \end{aligned}$$

- **Fourier Transform of product of the original $x[n]$ and r^{-n} .**
- If $r=1$, then FT of $x[n]$.

For unit circle in z -plane, we can evaluate $X[z]$ at points on the unit circle:

➡ Fourier Transform for $0 \leq \omega \leq \pi$

❖ Region of Convergence (ROC)

- FT and z-transform do not converge for all sequences.
- For any seq., the set of values z for which the z-transform power converge is called "*the region of convergence* (ROC)".

If absolutely summable, FT converges to ~~~~.

$$|X(z)| < \infty \quad \text{if} \quad \sum_{n=-\infty}^{\infty} |x[n]| |z|^{-n} < \infty$$

From the above, the convergence depends on $|z|$!!!

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n} \quad : \text{Power series or Laurent series}$$

❖ Region of Convergence (ROC)

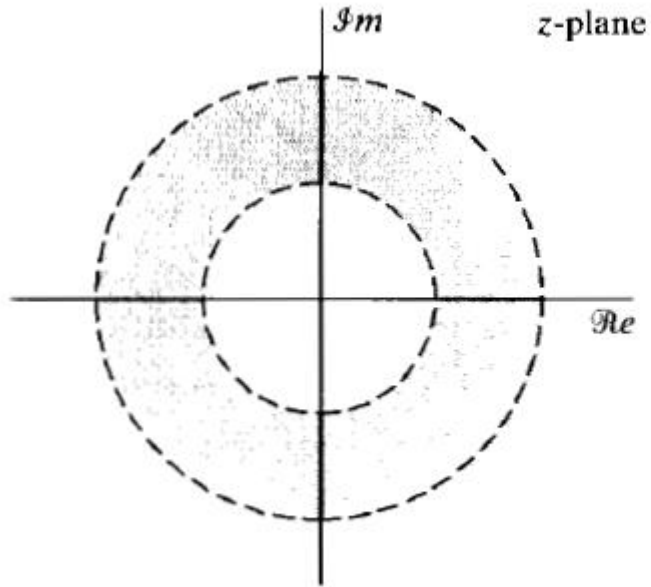


Figure 3.2 The region of convergence (ROC) as a ring in the z-plane. For specific cases, the inner boundary can extend inward to the origin, and the ROC becomes a disc. For other cases, the outer boundary can extend outward to infinity.

❖ Poles and Zeros

- In case of z-transform,

$$X(z) = \frac{P(z)}{Q(z)}$$

where $P(z)$ and $Q(z)$ are polynomials in z .

- **Poles:** z value of $X(z) = 0$.
- **Zeros:** z value of $X(z) = \infty$.

Important relationships exists between the locations of poles of $X(z)$ and The ROC of the z-transform.

z-Transform (6)

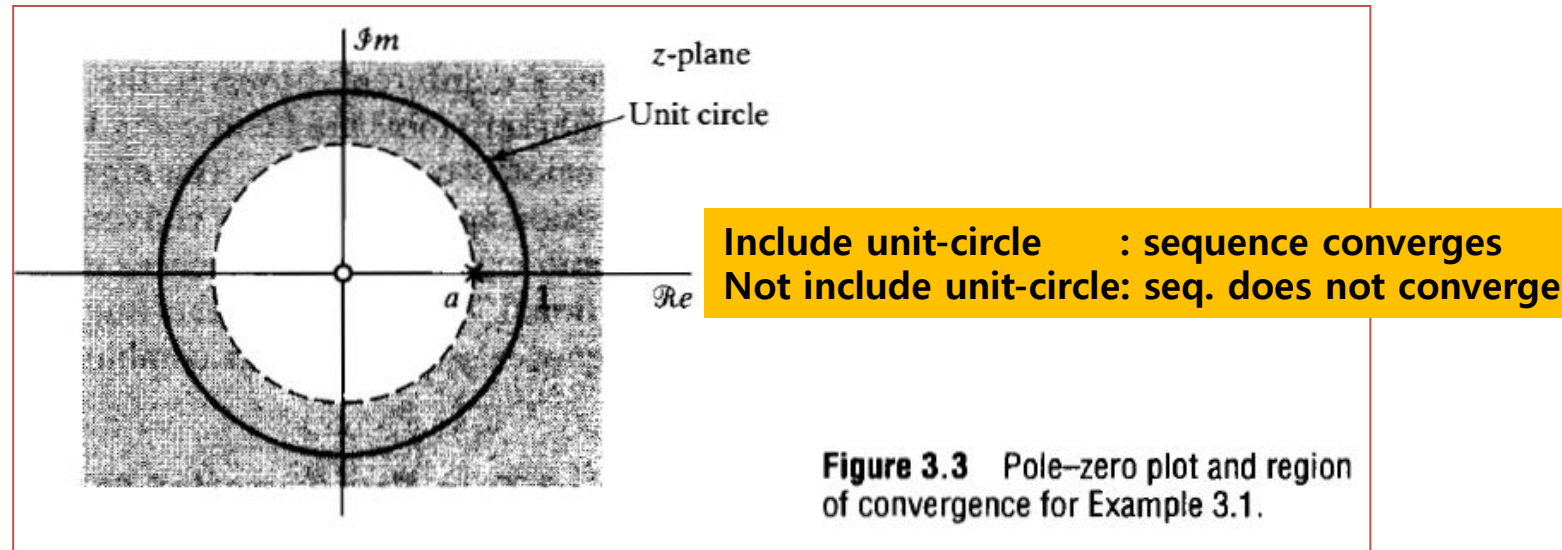
- Ex 3.1) signal $x[n] = a^n u[n]$

$$X(z) = \sum_{n=0}^{\infty} (az^{-1})^n$$

For convergence,

$$X(z) = \sum_{n=0}^{\infty} (az^{-1})^n < \infty$$

Then, $X(z) = \frac{1}{1 - az^{-1}}$ for $|az^{-1}| < 1 (|z| > |a|)$.



- Ex 3.3) sum of two exponential sequences

$$x[n] = \left(\frac{1}{2}\right)^n u[n] + \left(-\frac{1}{3}\right)^n u[n]$$

$$X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}} + \frac{1}{1 + \frac{1}{3}z^{-1}}$$

For convergence of $X(z)$, $\left|\frac{1}{2}z^{-1}\right| < 1$ and $\left|-\frac{1}{3}z^{-1}\right| < 1$

$$\therefore |z| > \frac{1}{2}$$

(ROC can be plotted as the next...!!!)

z-Transform (8)

- ROC of the previous slide

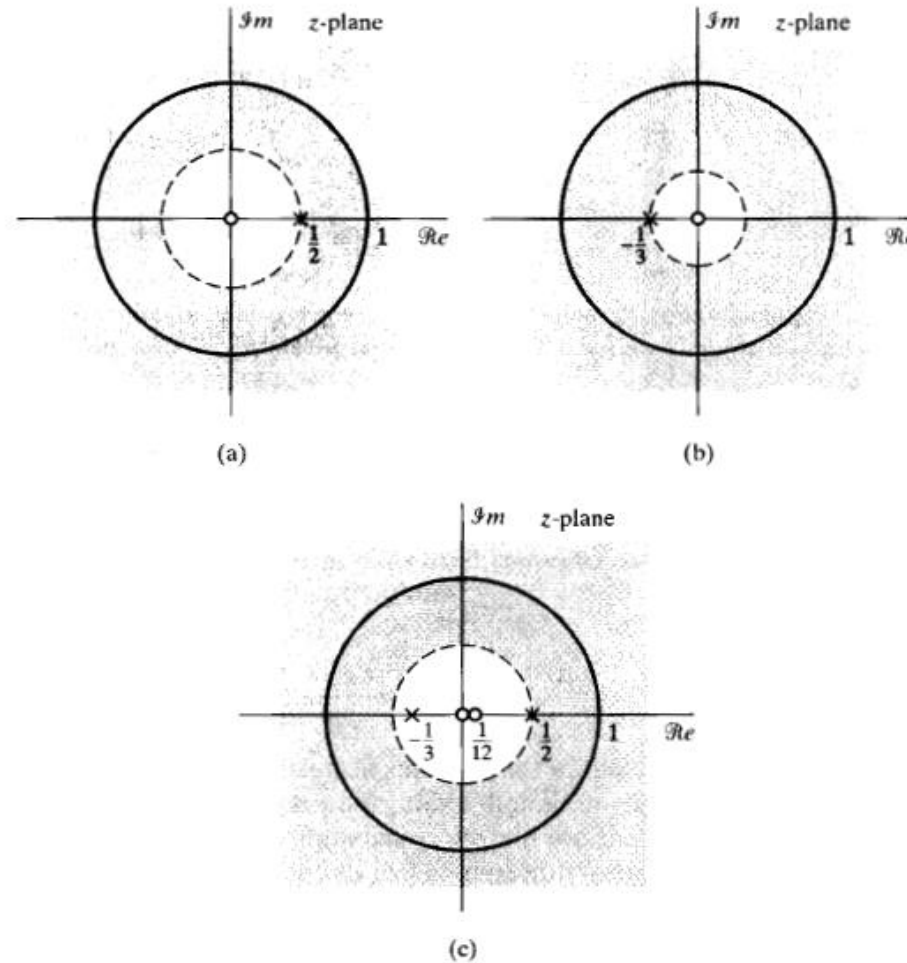
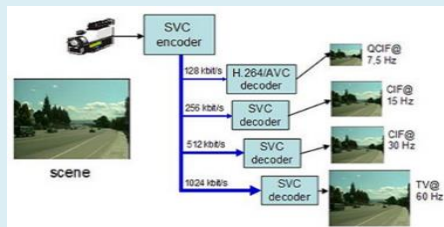
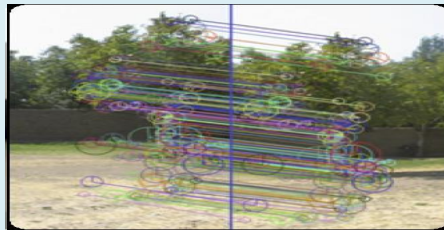


Figure 3.5 Pole-zero plot and region of convergence for the individual terms and the sum of terms in Examples 3.3 and 3.4. (a) $1/(1 - \frac{1}{2}z^{-1})$, $|z| > \frac{1}{2}$. (b) $1/(1 + \frac{1}{3}z^{-1})$, $|z| > \frac{1}{3}$. (c) $1/(1 - \frac{1}{2}z^{-1}) + 1/(1 + \frac{1}{3}z^{-1})$, $|z| > \frac{1}{2}$.

TABLE 3.1 SOME COMMON z-TRANSFORM PAIRS

Sequence	Transform	ROC
1. $\delta[n]$	1	All z
2. $u[n]$	$\frac{1}{1 - z^{-1}}$	$ z > 1$
3. $-u[-n - 1]$	$\frac{1}{1 - z^{-1}}$	$ z < 1$
4. $\delta[n - m]$	z^{-m}	All z except 0 (if $m > 0$) or ∞ (if $m < 0$)
5. $a^n u[n]$	$\frac{1}{1 - az^{-1}}$	$ z > a $
6. $-a^n u[-n - 1]$	$\frac{1}{1 - az^{-1}}$	$ z < a $
7. $na^n u[n]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z > a $
8. $-na^n u[-n - 1]$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z < a $
9. $[\cos \omega_0 n] u[n]$	$\frac{1 - [\cos \omega_0] z^{-1}}{1 - [2 \cos \omega_0] z^{-1} + z^{-2}}$	$ z > 1$
10. $[\sin \omega_0 n] u[n]$	$\frac{[\sin \omega_0] z^{-1}}{1 - [2 \cos \omega_0] z^{-1} + z^{-2}}$	$ z > 1$
11. $[r^n \cos \omega_0 n] u[n]$	$\frac{1 - [r \cos \omega_0] z^{-1}}{1 - [2r \cos \omega_0] z^{-1} + r^2 z^{-2}}$	$ z > r$
12. $[r^n \sin \omega_0 n] u[n]$	$\frac{[r \sin \omega_0] z^{-1}}{1 - [2r \cos \omega_0] z^{-1} + r^2 z^{-2}}$	$ z > r$
13. $\begin{cases} a^n, & 0 \leq n \leq N - 1, \\ 0, & \text{otherwise} \end{cases}$	$\frac{1 - a^N z^{-N}}{1 - az^{-1}}$	$ z > 0$



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Properties of the ROC for z-transform (1)

❖ Properties of the ROC

- *Usually, the properties of the ROC depend on the nature of the signal....!!!*
- *Property 1 ~ 8*

The Inverse z-Transform (1)

❖ Inverse z-transform

- Is represented as complex contour integral:

$$x[n] = \frac{1}{2\pi j} \oint_C X(z) z^{n-1} dz,$$

where C represents a closed contour within the ROC of the z-transform.

❖ Methods for Inverse z-transform

- Inspection Method
 - By inspection certain transform pairs.
 - For example, $a^n u[n] \longleftrightarrow \frac{1}{1 - az^{-1}}$, where $|z| > |a|$.
- Then the inverse of $X(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}$ equals ?

The Inverse z-Transform (2)

- Partial Fraction Expansion

- Sometimes, $X(z)$ may not be given explicitly in an available table.

Let us assume that $X(z)$ = a ratio of polynomials like

$$\begin{aligned} X(z) &= \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}} \\ &= \frac{z^N \sum_{k=0}^M b_k z^{-k}}{z^M \sum_{k=0}^N a_k z^{-k}} \end{aligned}$$

- If $M > N$, then $(M-N)$ poles at $z=0$.
- If $N > M$, then $(N-M)$ zeros at $z=0$.

- Also,

$$X(z) = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

where c_k = nonzero zeros of $X(z)$, d_k = nonzero poles of $X(z)$.

The Inverse z-Transform(3)

- Case 1) If $M < N$ and poles are all the 1-st order, then

$$X(z) = \sum_{k=1}^N \frac{A_k}{1 - d_k z^{-1}}$$

By multiplying $(1 - d_k z^{-1})$ and $z = d_k$,

$$\therefore A_k = (1 - d_k z^{-1})X(z)|_{z=d_k}$$

- Ex 3.9) 2-nd order z-transform

$$X(z) = \frac{1}{(1 - \frac{1}{4}z^{-1})(1 - \frac{1}{2}z^{-1})} \quad |z| > \frac{1}{2}$$

Let

$$X(z) = \frac{A_1}{(1 - \frac{1}{4}z^{-1})} + \frac{A_2}{(1 - \frac{1}{2}z^{-1})} \quad |z| > \frac{1}{2}$$

The Inverse z-Transform(4)

After computing A_1 and A_2 ,

$$X(z) = \frac{-1}{(1 - \frac{1}{4}z^{-1})} + \frac{2}{(1 - \frac{1}{2}z^{-1})} \quad |z| > \frac{1}{2}$$

From Table 3.1 and the linearity of z-transform,

$$\therefore x[n] = 2\left(\frac{1}{2}\right)^n u[n] - \left(\frac{1}{4}\right)^n u[n].$$

- In case of $M \geq N$, the complete partial fraction expression would be

$$X(z) = \sum_{r=0}^{M-N} B_r z^{-r} + \sum_{k=0}^N \frac{A_k}{1 - d_k z^{-1}}$$

Normal parts

Fraction parts

The Inverse z-Transform(5)

- Ex 3.10) Inverse by partial fraction

$$X(z) = \frac{1 + 2z^{-1} + z^{-2}}{(1 - \frac{1}{2}z^{-1})(1 - z^{-1})} \quad |z| > \frac{1}{2}$$

$$\text{Let } X(z) = B_0 + \frac{A_1}{(1 - \frac{1}{2}z^{-1})} + \frac{A_2}{(1 - 2z^{-1})} \quad |z| > \frac{1}{2}$$

$$\therefore B_0 = 2 \quad (\text{by direct division})$$

Then,

$$X(z) = 2 + \frac{-1 + 5z^{-1}}{(1 - \frac{1}{2}z^{-1})(1 - z^{-1})}$$

Now for coefficient A_1 and A_2 , by $A_k = (1 - d_k z^{-1})X(z)|_{z=d_k}$

$$A_1 = (2 + \frac{-1 + 5z^{-1}}{(1 - \frac{1}{2}z^{-1})(1 - z^{-1})})(1 - \frac{1}{2}z^{-1})|_{z=\frac{1}{2}} = -9$$

$$A_2 = 8$$

Therefore,

The Inverse z-Transform(6)

From Table 3.1,

$$\therefore X(z) = 2 - \frac{9}{(1 - \frac{1}{2}z^{-1})} + \frac{8}{(1 - z^{-1})}$$

$$\begin{aligned} 2 &\longleftrightarrow 2\delta[n], \\ \frac{1}{1 - \frac{1}{2}z^{-1}} &\longleftrightarrow \left(\frac{1}{2}\right)^n u[n], \\ \frac{1}{1 - z^{-1}} &\longleftrightarrow u[n], \end{aligned}$$

From the linearity of the z-transform,

$$\therefore x[n] = 2\delta[n] - 9\left(\frac{1}{2}\right)^n u[n] + 8u[n].$$

The Inverse z-Transform(7)

- Power Series Expansion

$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} x[n]z^{-n} \\ &= \cdots + x[-2]z^2 + x[-1]z + x[0] + x[1]z^{-1} + \cdots \end{aligned}$$

Ex 3.11) Finite-length Series

Suppose $X(z) = z^2(1 - \frac{1}{2}z^{-1})(1 + z^{-1})(1 - z^{-1}),$



Power series evaluation and order matching

$$\therefore x[n] = \delta[n] - \frac{1}{2}\delta[n] - \delta[n] + \frac{1}{2}\delta[n - 1].$$

It is very useful when $X(z)$ is the ratio of polynomials...!!!!

z-transform Properties (1)

❖ For LTI system,

$$x_1[n] \longleftrightarrow X_1(z), \text{ ROC} = R_{x_1}$$

$$x_2[n] \longleftrightarrow X_2(z), \text{ ROC} = R_{x_2}$$

- Linearity

$$ax_1[n] + bx_2[n] \longleftrightarrow aX_1(z) + bX_2(z), \text{ ROC contains } R_{x_1} \cap R_{x_2}.$$

- Time shifting

$$x[n - n_0] \longleftrightarrow Z^{-n_0} X(z)$$

- Multiplication by an Exponential Sequence

$$z_0^n x[n] \longleftrightarrow X(z/z_0), \text{ ROC} = |z_0| R_x.$$

- Differentiation of $X(z)$

$$nx[n] \longleftrightarrow -z \frac{X(z)}{dz}, \text{ ROC} = R_x.$$

How to prove????

z-transform Properties (2)

- Conjugation of a Complex Sequence

$$x^*[n] \longleftrightarrow X^*(z^*), \quad ROC = R_x$$

- Time-Reversal

$$x^*[-n] \longleftrightarrow X^*(1/z^*), \quad ROC = \frac{1}{R_x}$$

If $x[n]$ is real, then

$$x[-n] \longleftrightarrow X(1/z), \quad ROC = \frac{1}{R_x}$$

- Convolution of Sequence

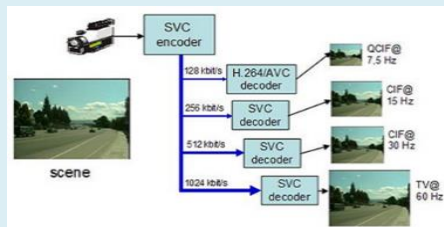
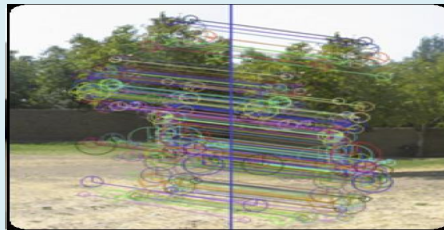
$$x_1[n] * x_2[n] \longleftrightarrow X_1(z)X_2(z), \quad \text{ROC contains } R_{x_1} \cap R_{x_2}.$$

- Some z-transform properties
(Refer to Table 3.2....!!!)

z-transform Properties (3)

TABLE 3.2 SOME z-TRANSFORM PROPERTIES

Section Reference	Sequence	Transform	ROC
	$x[n]$	$X(z)$	R_x
	$x_1[n]$	$X_1(z)$	R_{x_1}
	$x_2[n]$	$X_2(z)$	R_{x_2}
3.4.1	$ax_1[n] + bx_2[n]$	$aX_1(z) + bX_2(z)$	Contains $R_{x_1} \cap R_{x_2}$
3.4.2	$x[n - n_0]$	$z^{-n_0} X(z)$	R_x , except for the possible addition or deletion of the origin or ∞
3.4.3	$z_0^n x[n]$	$X(z/z_0)$	$ z_0 R_x$
3.4.4	$nx[n]$	$-z \frac{dX(z)}{dz}$	R_x , except for the possible addition or deletion of the origin or ∞
3.4.5	$x^*[n]$	$X^*(z^*)$	R_x
	$\text{Re}\{x[n]\}$	$\frac{1}{2}[X(z) + X^*(z^*)]$	Contains R_x
	$\text{Im}\{x[n]\}$	$\frac{1}{2j}[X(z) - X^*(z^*)]$	Contains R_x
3.4.6	$x^*[-n]$	$X^*(1/z^*)$	$1/R_x$
3.4.7	$x_1[n] * x_2[n]$	$X_1(z)X_2(z)$	Contains $R_{x_1} \cap R_{x_2}$
3.4.8	Initial-value theorem: $x[n] = 0, \quad n < 0 \quad \lim_{z \rightarrow \infty} X(z) = x[0]$		

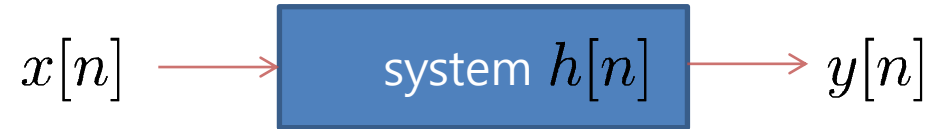


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z-Transform and LTI systems (1)

❖ LTI systems



$$y[n] = x[n] * h[n]$$

$$Y(z) = H(z)X(z) \quad \therefore \quad H(z) = \frac{Y(z)}{X(z)} : \text{system function (transfer function) of LTI system}$$

Ex 3.20) Convolution of ~

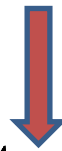
$$x[n] = Au[n], \quad h[n] = a^n u[n]$$
$$H(z) = \sum_{n=0}^{\infty} a^n z^{-n} = \frac{1}{1 - az^{-1}}, \quad |z| > |a|. \quad X(z) = \frac{A}{1 - z^{-1}}, \quad |z| > 1.$$

$$Y(z) = H(z)X(z) =$$

z-Transform and LTI systems (2)

$$= \frac{A}{(1 - az^{-1})(1 - z^{-1})},$$

$$= \frac{A}{1 - a} \left(\frac{1}{1 - z^{-1}} - \frac{a}{1 - az^{-1}} \right), \quad |z| > 1$$



Inverse of LTI system

$$\therefore y[n] = \frac{A}{1 - a} (1 - a^{n+1}) u[n].$$

❖ Some cases, LTI system function:

$$H(z) = \frac{\sum_{k=0}^M b_k z^{-k}}{\sum_{k=0}^N a_k z^{-k}} \quad \sum_{k=0}^M b_k x[n - k] = \sum_{k=0}^N a_k y[n - k]$$

The Unilateral z-Transform

❖ Unilateral or One-sided z-Transform

$$\mathcal{X}(z) = \sum_{n=0}^{\infty} x[n]z^{-n}$$

(Plz, read the details in pp. 164-166)

HW#2

- ❖ In textbook, Chap. 3,
 - Prob. 3.4, 3.7, 3.9, 3.17, 3.29
- ❖ Due day: ~ to the next week.

Thank you for your attention!!!
QnA

<http://ivpl.sookmyung.ac.kr>